

# Low-power radio design for the IoT

## Exercise 5 (25.03.2021)

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### Problem 1 Common Source Analysis

The circuit in the Figure below is a single MOSFET Common Source (CS) Amplifier.

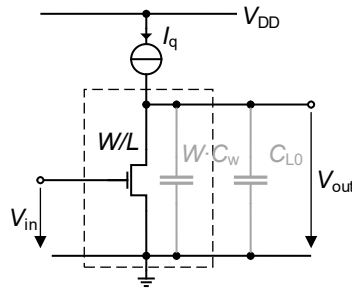


Figure 1: Common Source Amplifier.

- Draw the small-signal equivalent circuit of the CS. Consider the bias current source to be ideal. (*Hint*: split the gate capacitance,  $C_G$ , into the intrinsic and the extrinsic parts,  $C_{Gi}$  and  $C_{Ge}$ , and the load capacitance,  $C_L$ , into a constant capacitance and a self-loading capacitance per unit width,  $C_{L0}$  and  $C_w$ , respectively).
- Derive the expression of the transit frequency,  $\omega_t$ , by using the small signal equivalent circuit.
- Express  $\omega_t$  in terms of the inversion coefficient,  $IC$ . (*Hint*: introduce the specific frequency,  $\omega_{spec} \triangleq \frac{2\mu U_T}{L^2}$ , and neglect the velocity saturation ( $\lambda_c = 0$ )).
- Derive  $\omega_{tspec}$ , the transit frequency for  $IC = 1$  and WI.
- Derive the expression for the unity-gain frequency,  $\omega_u$ . (*Hint*: neglect the output conductance of the transistor ( $g_{ds} = 0$ )).
- Express  $\omega_u$  in terms of the normalized gain-bandwidth (GBW) for a squared transistor biased in WI,  $\omega_L \triangleq \frac{I_{spec\Box}}{nU_T C_L}$ , and the normalized source transconductance,  $g_{ms}$ .

### Problem 2 Common Source Design

Once the self loading capacitance per unit width is taken into account, the formulas for the  $I_D$  and  $W/L$  normalized to  $\Omega \triangleq \frac{\omega_u}{\omega_L}$  are

$$i_b \triangleq \frac{I_D}{I_{spec\Box}} \cdot \frac{1}{\Omega} = \frac{IC}{g_{ms} - \Theta}; \quad (1)$$

$$AR \triangleq \frac{W}{L} \cdot \frac{1}{\Omega} = \frac{1}{g_{ms} - \Theta},$$

where  $\Theta$  is equal to  $\frac{C_{uL}}{C_{L0}} \cdot \frac{\omega_u}{\omega_L} = \frac{\omega_u}{\omega_{tspec}}$ . Design the CS amplifier, shown in Fig. 1, for the following specifications at room temperature:

$$f_u = 18 \text{ MHz} \quad C_{L0} = 60 \text{ fF} \quad V_{DD} = 1.8 \text{ V} \quad L = 40 \text{ nm} \quad C_w = 0.450 \text{ fF/nm}, \quad (2)$$

and by assuming the following values for the technology parameters

$$I_{spec\Box} = 950 \text{ nA} \quad n = 1.5 \quad V_{T0} = 455 \text{ mV} \quad \lambda_c = 0.4875 \quad L_{sat} = 19.5 \text{ nm}. \quad (3)$$

- Find the  $IC_{opt}$ , the value of the inversion coefficient for which the bias current is minimum. Assume no velocity saturation.
- Based on the  $IC_{opt}$ , find the values of the bias current,  $I_q$ , and the transistor aspect ratio,  $W/L$ , to achieve the specified gain-bandwidth,  $w_u$ .