

Information Theory & Coding

Oct. 19th 2020.

So far:

- Source Coding -
- Prefix-free, unig. Decodable; Kraft Sum, etc.
- Huffman Codes
- Entropy as a lower bound to the # of bits per letter
- LZ algo as a "universal" method
- $H(U)$, $H(UV)$, $H(U_1 \dots U_n)$
 $H(U|V)$, chain rules, entropy rate
$$H(\{U_i\}) = \lim_{n \rightarrow \infty} \frac{1}{n} H(U_1 \dots U_n)$$
- mutual inf. $I(U; V) = H(U) - H(U|V)$
conditional versions, chain rules, etc..
- Data processing theorem
 $U - V - W \Rightarrow I(U; V) \geq I(U; W)$

- Fano's Ineq. $u, v \in \mathcal{U}$

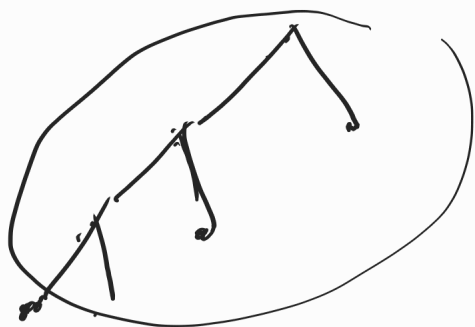
$$H(u|v) \leq h_2(p) + p h_2(u|v)$$

$$p = P(u \neq v), \quad h_2(p) = p \log_2 \frac{1}{p} + (1-p) \log_2 \frac{1}{1-p}$$

Recall $H(u) \sim$ guessing effort to learn

the value of the RV U

[Analogy between prefix-free codes & the game of 20 questions]



$$H(u|v) := \sum_v H(u|v=v) p_v(v)$$

$$H(u|v) \leq H(u).$$

$$I(u; v) = H(u) - H(u|v)$$

$$I(u; v) = D(P_{uv} \| P_u P_v)$$

Given (P_{uv}) let $(\tilde{u}_i, \tilde{v}_i)$ be iid RVs
 with distribution $\tilde{P}_{uv} = P_u \cdot P_v$.

$$\left(\underline{P_{uv}}, \text{ let } \underline{P_u(u)} = \sum_v P_{uv}(u, v) \right.$$

$$\underline{P_v(v)} = \sum_u P_{uv}(u, v)$$

$\tilde{P}_{uv}(u, v) = P_u(u)P_v(v)$ is also a distribution)

$$I(u; v) = D(P_{uv} \parallel \tilde{P}_{uv})$$

$$P_r(\underbrace{((\tilde{u}_1, \tilde{v}_1), (\tilde{u}_2, \tilde{v}_2), \dots, (\tilde{u}_n, \tilde{v}_n))}_{\text{is}} \in \underbrace{T(n, \varepsilon, P_{uv})}_{\text{is}})$$

$$\approx 2^{-n(D(P \parallel \tilde{P}) + o(\varepsilon))}$$

$$\approx 2^{-n \underline{I(u; v)}}$$

Data Transmission

Setup

A random variable U .

A:

U

B:



Channel is described by a probability kernel $P(y|x)$, for each $x \in \mathcal{X}$, $\sum_{y \in \mathcal{Y}} P(y|x) = 1$
 ≥ 0

$p(u, x_1, \dots, x_n, y_1, \dots, y_n)$

$$= P(U=u) P(X_1=x_1 | U=u) P(Y_1=y_1 | X_1=x_1, U=u)$$

$$P(X_2=x_2 | U=u, X_1=x_1, Y_1=y_1) P(Y_2=y_2 | X_2=x_2, X_1=x_1, Y_1=y_1, U=u)$$

$$= P(u=u) \prod_{i=1}^n P(X_i=x_i | U=u, (X_1, \dots, X_{i-1})=(x_1, \dots, x_{i-1}), (Y_1, \dots, Y_{i-1})=(y_1, \dots, y_{i-1}))$$

$P(Y_i=y_i | \dots, X_i=x_i)$

always true

Def:

A channel is called memoryless if
 a stationary

$$\left(\Pr(Y_i = y_i \mid U = u, \underbrace{(X_1, \dots, X_{i-1}) = (x_1, \dots, x_{i-1}), (Y_1, \dots, Y_{i-1}) = (y_1, \dots, y_{i-1})}_{X_i = x_i}) \right)$$

$$= \underline{p(y_i | x_i)}$$

Def.

• A communication system is called "without feedback" if

$$\Pr(X_i = x_i \mid U = u, \underbrace{(X_1, \dots, X_{i-1}) = (x_1, \dots, x_{i-1}), (Y_1, \dots, Y_{i-1}) = (y_1, \dots, y_{i-1})})$$

$$= \Pr(X_i = x_i \mid (U, X_1, \dots, X_{i-1}) = (u, x_1, \dots, x_{i-1}))$$

So: if we have a stationary, memoryless channel & system w/o feedback, then

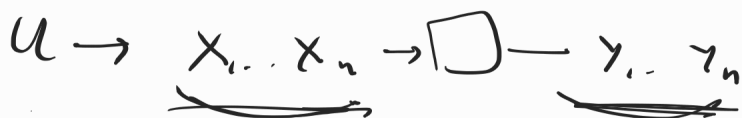
$$\Pr(U = u, X^n = x^n, Y^n = y^n) = p(u) \prod_{i=1}^n p(x_i | x^{i-1}, u) p(y_i | x_i)$$

$$[a^n = a_1 \dots a_n] = p(u) p(x^n | u) p(y^n | x^n)$$

where $p(y^n | x^n) = \prod_{i=1}^n p(y_i | x_i)$

We see that in such a system

$$\left(U - X^n - Y^n \right) \&$$

$$p(y^n | x^n) = \prod_{i=1}^n p(y_i | x_i)$$


no hidden connections $\equiv U - X^n - Y^n$

In particular, in such a system

$$\underline{H(Y^n | X^n)} = E \log \frac{1}{p(Y^n | X^n)}$$

$$= E \left(\log \frac{1}{\prod_{i=1}^n p(Y_i | X_i)} \right)$$

$$= \sum_{i=1}^n E \left(\log \frac{1}{p(Y_i | X_i)} \right) = \sum_{i=1}^n H(Y_i | X_i)$$

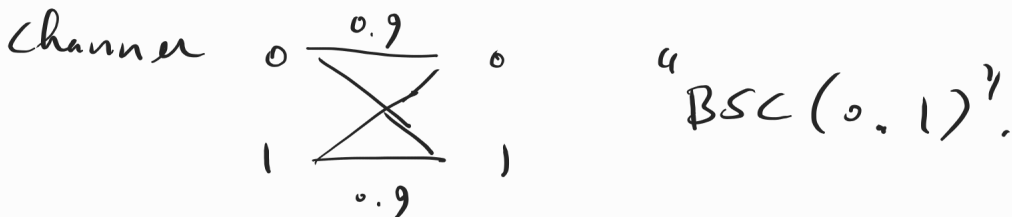
Example $\mathcal{U} = \{1, 2, 3\}$

$$P_U \quad \frac{1}{3} \quad \frac{1}{3} \quad \frac{1}{3}$$

when $u=1$ $X_1, X_2 = \begin{cases} \underline{0} \textcircled{0} & \frac{1}{2} \\ \underline{0} \textcircled{1} & \frac{1}{2} \end{cases}$

$u=2$ $X_1, X_2 = \underline{0} \textcircled{0}$

$u=3$ $X_1, X_2 = 11$



$$P_r(u=1, X^2 = \underline{00}, Y^2 = 01) = \frac{1}{3} \cdot \frac{1}{2} (0.9)(0.1)$$

$$P_r(u=2, X^2 = \underline{00}, Y^2 = 01) = \frac{1}{3} \cdot 1 (0.9)(0.1)$$

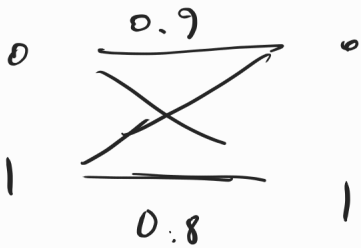
$$P_r(u=3, X^2 = 11, Y^2 = 01) = \frac{1}{3} \cdot 1 (0.9)(0.1)$$

$$H(\gamma_1 | X_1) = ?$$

$$\sum_{x_1} P(X_1 = x_1) \underbrace{\sum_{\gamma_1} P(\gamma_1 = \gamma_1 | X_1 = x_1) \ln \frac{1}{P(\gamma_1 = \gamma_1 | X_1 = x_1)}}_{h_2(0.1)}$$

$$= h_2(0.1)$$

If we keep u, X^2 same & change



$$H(\gamma_1 | X_1) = \sum_{x_1} P(X_1 = x_1) \underbrace{\sum_{\gamma_1} P(\gamma_1 = \gamma_1 | X_1 = x_1) \ln \frac{1}{P(\gamma_1 = \gamma_1 | X_1 = x_1)}}_{\begin{cases} h_2(0.1) & \text{if } x_1 = 0 \\ h_2(0.2) & \text{if } x_1 = 1 \end{cases}}$$

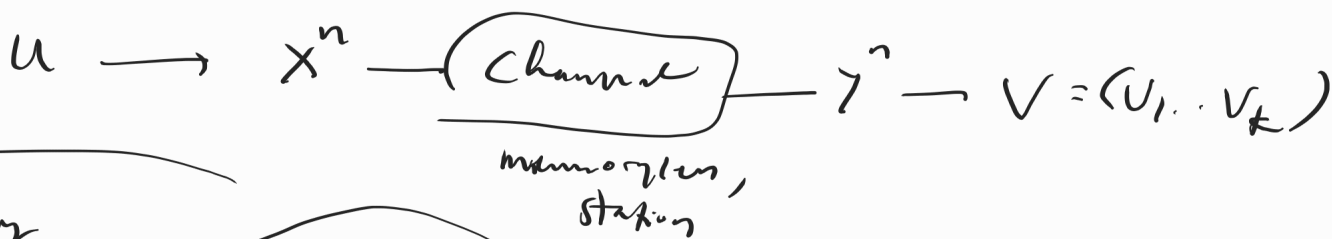
$$= \underbrace{P(X_1 = 0)}_{\frac{2}{3}} h_2(0.1) + \underbrace{P(X_1 = 1)}_{\frac{1}{3}} h_2(0.2)$$

$$H(\gamma_2 | X_2) = \underbrace{P(X_2 = 0)}_{\frac{1}{2}} h_2(0.1) + \underbrace{P(X_2 = 1)}_{\frac{1}{2}} h_2(0.2)$$

$$H(\gamma^2 | X^2) = H(\gamma_1 | X_1) + H(\gamma_2 | X_2) \approx \dots$$

Suppose we have a communication system of the type we have above

$$U = (u_1, \dots, u_k)$$



may we want $P_e(U \neq V)$ small $\rightarrow E[\max_i \mathbb{1}\{u_i \neq v_i\}]$

we may want $\frac{1}{k} \sum_{i=1}^k P_e(u_i \neq v_i) = E\left[\frac{\# \text{ of wrong } i\text{'s}}{k}\right]$

$\mathbb{1}\{u_i \neq v_i\}$ $\rightarrow \frac{1}{k} \sum_{i=1}^k \mathbb{1}\{u_i \neq v_i\}$

Fano's Inequality revisited

$H(U|V) \leq h_2(p) + p \log(|\mathcal{U}| - 1)$

we want

$\frac{1}{k} H(U^k|V^k) \leq h_2(\bar{p}) + \bar{p} \log(|\mathcal{U}| - 1)$

with $\bar{p} = \frac{1}{k} \sum_{i=1}^k P_e(u_i \neq v_i)$

all u_i, v_i belong to \mathcal{U} .

Proof:

$$H(\underline{u}^k | \underline{v}^k) = \sum_{i=1}^k H(u_i | \underline{u}^{i-1} \underline{v}^k) \quad \text{chain rule}$$

$$\leq \sum_{i=1}^k H(u_i | v_i) \quad \text{cond. red. ent.}$$

$$\leq h_2(p_i) + \underbrace{(p_i)}_{\substack{p_i = P(u_i | \underline{v}^k) \\ \text{Fano.}}} \log_2(u_i - 1)$$

$$\therefore \frac{1}{k} H(\underline{u}^k | \underline{v}^k) \leq \frac{1}{k} \sum_i [\quad]$$

$$= \underbrace{\frac{1}{k} \sum_{i=1}^k h_2(p_i)} + \bar{p} \log_2(u-1)$$

$$\leq \underbrace{h_2(\bar{p})} + \bar{p} \log_2(u-1)$$

Concavity of $h_2(\cdot)$.

$h_2(p) = -p \log_2 p - (1-p) \log_2 (1-p)$ is concave.

$$\frac{\partial h_2(p)}{\partial p} = -\log_2 p - 1 + \log_2 (1-p) + 1$$

$$= \log_2 (1-p) - \log_2 p$$

$$\frac{\partial^2}{\partial p^2} h_2(p) = -\frac{1}{1-p} - \frac{1}{p} = -\frac{1}{p(1-p)} \leq 0.$$



Given a memoryless, stationary channel $p(y|x)$

compute $\underline{C} = \left(\max_{P_X} I(X; Y) \right)$ Then in

$$u^k - X^n \xrightarrow{ch} Y^n - v^k$$

$\bar{p} = \frac{1}{k} \sum P_{r_i}(u_i \neq v_i)$ satisfies

$$\underline{h_2(\bar{p}) + \bar{p} \log_2(1/\bar{p})} \geq \underline{\left[\frac{1}{k} H(u^k) - \frac{n}{k} C \right]}$$

Pf: By the Reverse Fano

$$h_2(\bar{p}) + \bar{p} \log_2(1/\bar{p}) \geq \frac{1}{k} H(u^k | v^k)$$

$$= \frac{1}{k} [H(u^k) - I(u^k; v^k)]$$

$$\geq \frac{1}{k} [H(u^k) - \underbrace{I(X^n; Y^n)}_{\text{Data proc.}}]$$

$$\geq \frac{1}{k} [H(u^k) - \underbrace{nC}]$$

$$\underbrace{u^k - X^n - Y^n - v^k}$$

because

$$\begin{aligned} I(X^n; Y^n) &= H(Y^n) - H(Y^n | X^n) \\ &= H(Y^n) - \sum_{i=1}^n H(Y_i | X_i) \\ &\leq \sum_{i=1}^n H(Y_i) - \end{aligned}$$

$$= \sum_{i=1}^n \underbrace{I(x_i; y_i)}_{\leq C} \leq nC. \quad //$$

Consequently:

if u_1, u_2, \dots is a stationary source we also

$$\text{have } \frac{1}{k} H(u_1 \dots u_k) = \frac{1}{k} \sum_{i=1}^k H(u_i | u^{i-1})$$

$$= \frac{1}{k} [H(u_1) + H(u_2 | u_1) + \dots + H(u_k | u^{k-1})]$$

$$= \frac{1}{k} (H(u_k) + H(u_{k-1} | u_{k-2}) + \dots + H(u_2 | u^1))$$

$$\geq \frac{1}{k} \cdot k H(u_k | u^{k-1}) \rightarrow$$

$$\geq H(u_i).$$

$$\Rightarrow \textcircled{p} \text{ satisfies } \underbrace{h_2(\bar{p}) + \bar{p} \log_2(1/\bar{p})} \geq \underbrace{H(\bar{p}) + \frac{n}{k} C}$$

$\frac{n}{k}$: # of times the channel is used per transmitted source letter.